

#### Features

- Supply Current: 950nA Maximum /Amplifier
- Stable 18kHz GBWP with 10mV/µs Slew Rate
- Offset Voltage: 1.5mV Maximum
- Ultra-low Vos TC: 0.5µV/°C
- Low Input Bias Current: 1pA Typical
- High 120dB Open-Loop Voltage Gain
- Unity Gain Stable for 1,000nF Capacitive Load
- Rail-to-Rail Input/Output Voltage Range
- Outputs Source and Sink 20mA of Load Current
- No Phase Reversal for Overdriven Inputs
- Ultra-low Single-Supply Operation Down to +1.8V
- -40°C to 125°C Operation Range
- Robust 6kV HBM and 1.5kV CDM ESD Rating
- Green, Popular Type Package

### **Applications**

- Handsets and Mobile Accessories
- Current Sensing
- Wireless Remote Sensors, Active RFID Readers
- Environment/Gas/Oxygen Sensors
- Threshold Detectors/Discriminators
- Low Power Filters
- Battery or Solar Powered Devices
- Sensor Network Powered by Energy Scavenging

1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps Description

The TP212x are ultra-low power, precision CMOS op-amps featuring a maximum supply current of 950nA per amplifier with an ultra-low typical input bias current of 1pA. Analog trim and calibration routine reduce input offset voltage to below 1.5mV, and the precision temperature compensation technique makes offset voltage temperature drift at  $0.5\mu$ V/°C, which allowing use of the TP212x in systems with high gain without creating excessively large output offset errors.

The TP212x are unity gain stable with 1,000nF capacitive load with a constant 18kHz GBWP, 10mV/ $\mu$ s slew rate, which make them appropriate for low frequency applications, such as battery current monitoring and sensor conditioning. The TP212x can operate from a single-supply voltage of +1.8V to +6.0V or a dual-supply voltage of ±0.9V to ±3.0V. Beyond the rails input and rail-to-rail output characteristics allow the full power-supply voltage to be used for signal range.

The combined features make the TP212x ideal choices for battery-powered applications because they minimize errors due to power supply voltage variations over the lifetime of the battery and maintain high CMRR even for a rail-to-rail input opamp. Mobile accessories, wireless remote sensing, backup battery sensors, and single-Li+ or 2-cell NiCd/Alkaline battery powered systems can benefit from the features of the TP212x op-amps.

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**TP2121 in Low Side Battery Current Sensor** 

#### **Ultra-low Supply Current Op-amps:**

Supply Current	0.3 µA	0.6 µA	4 µA
GBWP	10 kHz	18 kHz	150 kHz
Single	TP2111	TP2121	TP1511
Dual	TP2112	TP2122	TP1512
Quad	TP2114	TP2124	TP1514

## **Pin Configuration** (Top View)



#### **Order Information**

Model Name	Order Number	Package	MSL Level	Transport Media, Quantity	Marking Information
TP2121	TP2121-TR	5-Pin SOT23	MSL 3	Tape and Reel, 3,000	B2T
112121	TP2121-CR	5-Pin SC70	MSL 3	Tape and Reel, 3,000	B2C
TP2122	TP2122-SR	8-Pin SOIC	MSL 3	Tape and Reel, 4,000	B22S
172122	TP2122-VR	8-Pin MSOP	MSL 3	Tape and Reel, 3,000	B22V
TP2124	TP2124-SR	14-Pin SOIC	MSL 3	Tape and Reel, 2,500	B24S
112124	TP2124-TR	14-Pin TSSOP	MSL 3	Tape and Reel, 3,000	B24T

#### Absolute Maximum Ratings Note 1

Supply Voltage: V <sup>+</sup> – V <sup>-</sup>	6.5V
Input Voltage V <sup>-</sup> – 0.3 to V	+ + 0.3
Input Current: +IN, -INNote 2 ±	10mA
Output Short-Circuit Duration Note 3 Inc	lefinite

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The inputs are protected by ESD protection diodes to each power supply. If the input extends more than 500mV beyond the power supply, the input current should be limited to less than 10mA.

Note 3: A heat sink may be required to keep the junction temperature below the absolute maximum. This depends on the power supply voltage and how many amplifiers are shorted. Thermal resistance varies with the amount of PC board metal connected to the package. The specified values are for short traces connected to the leads.

## ESD, Electrostatic Discharge Protection

Symbol	Parameter	Condition	Minimum Level	Unit
HBM	Human Body Model ESD	ANSI/ESDA/JEDEC JS-001	6	kV
CDM	Charged Device Model ESD	ANSI/ESDA/JEDEC JS-002	1.5	kV

## **5V Electrical Characteristics**

The specifications are at  $T_A$  = 27°C.  $V_{DD}$  = 5V,  $V_{CM}$  =  $V_{OUT}$  =  $V_{DD}/2$ ,  $R_L$  = 100k $\Omega$ ,  $C_L$  =60pF.

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Vos	Input Offset Voltage	$V_{CM} = V_{DD}/2$ and $V_{CM} = GND$	-1.5	±0.1	+1.5	mV
Vos TC	Input Offset Voltage Drift			0.5		μV/°C
		T <sub>A</sub> =27 ∘C		1		, pA
В	Input Bias Current	T <sub>A</sub> =85 °C		20		рА
		T <sub>A</sub> =125 °C		100		pA
los	Input Offset Current			1		рА
Vn	Input Voltage Noise	f = 0.1Hz to 10Hz		6.5		μV <sub>P-P</sub>
en	Input Voltage Noise Density	f = 1kHz		170		nV/√Hz
R <sub>IN</sub>	Input Resistance			1		TΩ
<u>^</u>	lanut Canacitanaa	Differential		2.9		
CIN	Input Capacitance	Common Mode		5		pF
CMRR	Common Mode Rejection Ratio	V <sub>CM</sub> = 0.1V to 4.9V	60	100		dB
V <sub>CM</sub>	Common-mode Input Voltage Range		V0.3		V++0.3	V
PSRR	Power Supply Rejection Ratio	V <sub>DD</sub> = 1.8V to 5.5V	70	92		dB
A <sub>VOL</sub>	Open-Loop Large Signal Gain	$V_{OUT}$ = 0.5V to 4.5V, $R_{LOAD}$ = 100k $\Omega$	80	120		dB
Vol, Voh	Output Swing from Supply Rail	$R_{LOAD} = 100 k\Omega$		5		mV
Rout	Closed-Loop Output Impedance	G = 1, f = 1kHz, I <sub>OUT</sub> = 0		0.4		Ω
Ro	Open-Loop Output Impedance	f = 1kHz, Iout = 0		2.6		Ω
lsc	Output Short-Circuit Current	Sink or source current		20		mA
Vdd	Supply Voltage		1.8		6.0	V
la	Quiescent Current per Amplifier			600	950	nA
PM	Phase Margin	$R_{LOAD} = 100k\Omega$ , $C_{LOAD} = 60pF$		61		0
GM	Gain Margin	$R_{LOAD} = 100k\Omega$ , $C_{LOAD} = 60pF$		10		dB
GBWP	Gain-Bandwidth Product	f = 1kHz		18		kHz
	Settling Time, 1.5V to 3.5V, Unity	0.1%		0.25		
1	Gain	0.01%		0.253		
ts	Settling Time, 2.45V to 2.55V,	0.1%		0.035		ms
	Unity Gain	0.01%		0.038		
SR	Slew Rate	$A_V$ = 1, V <sub>OUT</sub> = 1.5V to 3.5V, C <sub>LOAD</sub> = 60pF, R <sub>LOAD</sub> = 100kΩ		10		mV/µs

Note 1: Full power bandwidth is calculated from the slew rate FPBW = SR/ $\pi \cdot V_{P-P.}$ 

## 1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps

# **Typical Performance Characteristics**



Small-Signal Step Response, 100mV Step

Large-Signal Step Response, 2V Step



**Open-Loop Gain and Phase** 



Input Voltage Noise Spectral Density



Phase Margin vs. CLOAD (Stable for Any CLOAD)



**Common-Mode Rejection Ratio** 



### **Typical Performance Characteristics**

Over-Shoot Voltage,  $C_{LOAD}$  = 40nF, Gain = +1,  $R_{FB}$ =100k $\Omega$ 



Over-Shoot % vs. C<sub>LOAD</sub>, Gain = +1, R<sub>FB</sub> = 1MΩ



Over-Shoot Voltage,  $C_{LOAD}$ =40nF, Gain= -1,  $R_{FB}$ =100k $\Omega$ 



**Power-Supply Rejection Ratio** 



Over-Shoot % vs.  $C_{LOAD}$ , Gain = -1,  $R_{FB}$  = 1M $\Omega$ 



V<sub>IN</sub> = -0.2V to 5.7V, No Phase Reversal



## 1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps

### **Typical Performance Characteristics**



**Quiescent Supply Current vs. Temperature** 

#### **Quiescent Supply Current vs. Supply Voltage**







**Open-Loop Gain vs. Temperature** 



Short-Circuit Current vs. Supply Voltage



#### Input Offset Voltage vs. Common Mode Input Voltage



### **Typical Performance Characteristics**



#### **Closed-Loop Output Impedance vs. Frequency**





#### **Pin Functions**

**–IN:** Inverting Input of the Amplifier. Voltage range of this pin can go from  $V^- - 0.3V$  to  $V^+ + 0.3V$ .

**+IN:** Non-Inverting Input of Amplifier. This pin has the same voltage range as –IN.

**V+ or +Vs:** Positive Power Supply. Typically the voltage is from 1.8V to 5.5V. Split supplies are possible as long as the voltage between V+ and V– is between 1.8V and 5.5V. A bypass capacitor of  $0.1\mu$ F as close to the part as possible should be used between power supply pins or between supply pins and ground.

N/C: No Connection.

**V**<sup>-</sup> or <sup>-</sup>**V**<sub>S</sub>: Negative Power Supply. It is normally tied to ground. It can also be tied to a voltage other than ground as long as the voltage between V<sup>+</sup> and V<sup>-</sup> is from 1.8V to 5.5V. If it is not connected to ground, bypass it with a capacitor of  $0.1\mu$ F as close to the part as possible.

**OUT:** Amplifier Output. The voltage range extends to within milli-volts of each supply rail.

#### Operation

The TP212x family input signal range extends beyond the negative and positive power supplies. The output can even extend all the way to the negative supply. The input stage is comprised of two CMOS differential amplifiers, a PMOS stage and NMOS stage that are active over different ranges of common mode input voltage. The Class-AB control buffer and output bias stage uses a proprietary compensation technique to take full advantage of the process technology to drive very high capacitive loads. This is evident from the transient over shoot measurement plots in the Typical Performance Characteristics.

## **Applications Information**

### Low Supply Voltage and Low Power Consumption

The TP212x family of operational amplifiers can operate with power supply voltages from 1.8V to 6.0V. Each amplifier draws only 600nA quiescent current. The low supply voltage capability and low supply current are ideal for portable applications demanding HIGH CAPACITIVE LOAD DRIVING CAPABILITY and CONSTANT WIDE BANDWIDTH. The TP212x family is optimized for wide bandwidth low power applications. They have an industry leading high GBWP to power ratio and are unity gain stable for 1,000nF capacitive load. When the load capacitance increases, the increased capacitance at the output pushed the non-dominant pole to lower frequency in the open loop frequency response, lowering the phase and gain margin. Higher gain configurations tend to have better capacitive drive capability than lower gain configurations due to lower closed loop bandwidth and hence higher phase margin.

#### Low Input Referred Noise

The TP212x family provides a low input referred noise density of 170 nV/  $\sqrt{Hz}$  at 1kHz. The voltage noise will grow slowly with the frequency in wideband range, and the input voltage noise is typically  $6.5\mu$ V<sub>P-P</sub> at the frequency of 0.1Hz to 10Hz.

#### Low Input Offset Voltage

The TP212x family has a low offset voltage of 1.5mV maximum which is essential for precision applications. The offset voltage is trimmed with a proprietary trim algorithm to ensure low offset voltage for precision signal processing requirement.

#### Ground Sensing and Rail to Rail Output

The TP212x family has excellent output drive capability, delivering over 10mA of output drive current. The output stage is a rail-to-rail topology that is capable of swinging to within 5mV of either rail. Since the inputs can go 300mV beyond either rail, the op-amp can easily perform 'true ground' sensing.

The maximum output current is a function of total supply voltage. As the supply voltage to the amplifier increases, the output current capability also increases. Attention must be paid to keep the junction temperature of the IC below 150°C when the output is in continuous short-circuit. The output of the amplifier has reverse-biased ESD diodes connected to each supply. The output should not be forced more than 0.5V beyond either supply, otherwise current will flow through these diodes.

#### ESD

The TP212x family has reverse-biased ESD protection diodes on all inputs and output. Input and out pins can not be biased more than 300mV beyond either supply rail.

### **Driving Large Capacitive Load**

The TP212x family of OPA is designed to drive large capacitive loads. Refer to Typical Performance Characteristics for "Phase Margin vs. Load Capacitance". As always, larger load capacitance decreases overall phase margin in a feedback system where internal frequency compensation is utilized. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed-loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in output step response. The unity-gain buffer (G = +1V/V) is the most sensitive to large capacitive loads.

When driving large capacitive loads with the TP212x OPA family (e.g., > 200 pF when G = +1V/V), a small series resistor at the output (R<sub>ISO</sub> in Figure 2) improves the feedback loop's phase margin and stability by making the output load resistive at higher frequencies.





### Power Supply Layout and Bypass

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The TP212x OPA's power supply pin (V<sub>DD</sub> for single-supply) should have a local bypass capacitor (i.e.,  $0.01\mu$ F to  $0.1\mu$ F) within 2mm for good high frequency performance. It can also use a bulk capacitor (i.e.,  $1\mu$ F or larger) within 100mm to provide large, slow currents. This bulk capacitor can be shared with other analog parts.

Ground layout improves performance by decreasing the amount of stray capacitance and noise at the OPA's inputs and outputs. To decrease stray capacitance, minimize PC board lengths and resistor leads, and place external components as close to the op amps' pins as possible.

#### **Proper Board Layout**

To ensure optimum performance at the PCB level, care must be taken in the design of the board layout. To avoid leakage currents, the surface of the board should be kept clean and free of moisture. Coating the surface creates a barrier to moisture accumulation and helps reduce parasitic resistance on the board.

Keeping supply traces short and properly bypassing the power supplies minimizes power supply disturbances due to output current variation, such as when driving an ac signal into a heavy load. Bypass capacitors should be connected as closely as possible to the device supply pins. Stray capacitances are a concern at the outputs and the inputs of the amplifier. It is recommended that signal traces be kept at least 5mm from supply lines to minimize coupling.

A variation in temperature across the PCB can cause a mismatch in the Seebeck voltages at solder joints and other points where dissimilar metals are in contact, resulting in thermal voltage errors. To minimize these thermocouple effects, orient resistors so heat sources warm both ends equally. Input signal paths should contain matching numbers and types of components, where possible to match the number and type of thermocouple junctions. For example, dummy components such as zero value resistors can be used to match real resistors in the opposite input path. Matching components should be located in close proximity and should be oriented in the same manner. Ensure leads are of equal length so that thermal conduction is in equilibrium. Keep heat sources on the PCB as far away from amplifier input circuitry as is practical.

The use of a ground plane is highly recommended. A ground plane reduces EMI noise and also helps to maintain a constant temperature across the circuit board.

#### BATTERY CURRENT SENSING

The Common Mode Input voltage Range of TP212x OPA series, which goes 0.3V beyond both supply rails, supports their use in high-side and low-side battery current sensing applications. The low quiescent current (600nA, typical) helps prolong battery life, and the rail-to-rail output supports detection of low currents.

The battery current (I<sub>DD</sub>) through the 10 $\Omega$  resistor causes its top terminal to be more negative than the bottom terminal. This keeps the Common Mode Input voltage below V<sub>DD</sub>, which is within its allowed range. The output of the OPA will also be blow V<sub>DD</sub>, within its Maximum Output Voltage Swing specification.



Figure 3

#### Instrumentation Amplifier

The TP212x OPA series is well suited for conditioning sensor signals in battery-powered applications. Figure 4 shows a two op-amp instrumentation amplifier, using the TP212x OPA.

The circuit works well for applications requiring rejection of Common Mode noise at higher gains. The reference voltage ( $V_{REF}$ ) is supplied by a low-impedance source. In single voltage supply applications,  $V_{REF}$  is typically  $V_{DD}/2$ .



#### Figure 4

## **Buffered Chemical Sensor (pH) Probe**

The TP212x OPA has input bias current in the pA range. This is ideal in buffering high impedance chemical sensors such as pH probe. As an example, the circuit in Figure 5 eliminates expansive low-leakage cables that that is required to connect pH probe to metering ICs such as ADC, AFE and/or MCU. A TP212x OPA and a lithium battery are housed in the probe assembly. A conventional low-cost coaxial cable can be used to carry OPA's output signal to subsequent ICs for pH reading.



ALL COMPONENTS CONTAJNED WITHIN THE pH PROBE



## Portable Gas Sensor Amplifier

Gas sensors are used in many different industrial and medical applications. Gas sensors generate a current that is proportional to the percentage of a particular gas concentration sensed in an air sample. This output current flows through a load resistor and the resultant voltage drop is amplified. Depending on the sensed gas and sensitivity of the sensor, the output current can be in the range of tens of microamperes to a few milli-amperes. Gas sensor datasheets often specify a recommended load resistor value or a range of load resistors from which to choose.

There are two main applications for oxygen sensors – applications which sense oxygen when it is abundantly present (that is, in air or near an oxygen tank) and those which detect traces of oxygen in parts-per-million concentration. In medical applications, oxygen sensors are used when air quality or oxygen delivered to a patient needs to be monitored. In fresh air, the concentration of oxygen is 20.9% and air samples containing less than 18% oxygen are considered dangerous. In industrial applications, oxygen sensors are used to detect the absence of oxygen; for example, vacuum-packaging of food products.

The circuit in Figure 6 illustrates a typical implementation used to amplify the output of an oxygen detector. With the components shown in the figure, the circuit consumes less than 600nA of supply current ensuring that small form-factor single- or button-cell batteries (exhibiting low mAh charge ratings) could last beyond

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the operating life of the oxygen sensor. The precision specifications of these amplifiers, such as their low offset voltage, low  $V_{OS}$  TC, low input bias current, high CMRR, and high PSRR are other factors which make these amplifiers excellent choices for this application.



Figure 6

# Tape and Reel Information





Order Number	Package	D1 (mm)	W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	W0 (mm)	Pin1 Quadrant
TP2121-CR	SOT353 (SC70-5)	178.0	12.3	2.4	2.5	1.2	4.0	8.0	Q3
TP2121-TR	SOT23-5	180.0	13.1	3.2	3.3	1.4	4.0	8.0	Q3
TP2122-SR	SOP8	330.0	17.6	6.4	5.4	2.1	8.0	12.0	Q1
TP2122-VR	MSOP8	330.0	17.6	5.2	3.3	1.5	8.0	12.0	Q1
TP2124-SR	SOP14	330.0	21.6	6.5	9.1	1.8	8.0	16.0	Q1
TP2124-TR	TSSOP14	330.0	17.6	6.8	5.5	1.4	8.0	12.0	Q1

# **Package Outline Dimensions**

## SOT353 (SC70-5)

Package Outline Dimensions		SC5	(SOT	۲353·	-5-A)
		(			
			ensions		nsions
	Symbol	MIN	llimeters MAX	MIN	MAX
	Α	0.850	1.100	0.033	0.043
	A1	0.000	0.100	0.000	0.004
	A2	0.800	1.000	0.031	0.039
	b	0.150	0.350	0.006	0.014
	с	0.110	0.230	0.004	0.009
	D	2.000	2.200	0.079	0.087
	E	2.150	2.450	0.085	0.096
	E1	1.150	1.350	0.045	0.053
NOTES	е	0.65	0 BSC	0.02	6 BSC
	L	0.260	0.460	0.010	
<ol> <li>Do not include mold flash or protrusion.</li> </ol>	L L				0.018

1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps

# SOT23-5

Package Outline Dimensions		S5	T(SC	T23-	-5-A)
		(			
	Symbol		ensions Ilimeters	Dime	
	- Cyrillool	1			nsions nches
		MIN	MAX		
	A	MIN 1.050		In Ir	nches
	A A1		MAX	In Ir MIN	MAX
		1.050	MAX 1.250	In Ir MIN 0.041	MAX 0.049
	A1	1.050 0.000	MAX 1.250 0.150	In Ir MIN 0.041 0.000	MAX 0.049 0.006
	A1 A2	1.050 0.000 1.000	MAX 1.250 0.150 1.200	In Ir MIN 0.041 0.000 0.039	MAX           0.049           0.006           0.047
	A1 A2 b	1.050 0.000 1.000 0.280	MAX 1.250 0.150 1.200 0.500	In Ir MIN 0.041 0.000 0.039 0.011	MAX           0.049           0.006           0.047
	A1 A2 b c	1.050 0.000 1.000 0.280 0.100	MAX 1.250 0.150 1.200 0.500 0.230	In Ir MIN 0.041 0.000 0.039 0.011 0.004	MAX           0.049           0.006           0.047           0.020           0.009
	A1 A2 b c D	1.050 0.000 1.000 0.280 0.100 2.820	MAX 1.250 0.150 1.200 0.500 0.230 3.020	In Ir MIN 0.041 0.000 0.039 0.011 0.004 0.111	MAX           0.049           0.006           0.047           0.020           0.009           0.119
NOTES	A1 A2 b c D E	1.050 0.000 1.000 0.280 0.100 2.820 2.600 1.500	MAX 1.250 0.150 1.200 0.500 0.230 3.020 3.000	In Ir MIN 0.041 0.000 0.039 0.011 0.004 0.111 0.102 0.059	MAX           0.049           0.006           0.047           0.020           0.009           0.119           0.118
NOTES 1. Do not include mold flash or protrusion.	A1 A2 b c D E E1	1.050 0.000 1.000 0.280 0.100 2.820 2.600 1.500	MAX 1.250 0.150 1.200 0.500 0.230 3.020 3.020 1.720	In Ir MIN 0.041 0.000 0.039 0.011 0.004 0.111 0.102 0.059	MAX           0.049           0.006           0.047           0.020           0.009           0.119           0.118           0.068

### SOP8



## 1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps

## MSOP8



### SOP14



# 1.8V, 600nA Nanopower, Rail-to-Rail Input/Output Op-amps

## TSSOP14



Date	Revision	Notes
2023-04-07	Rev.A.1	The following updates are all about the new datasheet formats or typo, the actual product remains unchanged. Updated to new format of package dimensions. Updated specification with test limit. Added MSL information. Added tape and reel information.

# **Revision History**

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