### MAX40000/MAX40001

# 1.7V, nanoPower Comparators with Built-In Reference

### **General Description**

The MAX40000/MAX40001 are tiny, single comparators with built-in voltage references that are ideal for a wide variety of portable electronics applications, such as cell phones, portable instruments, and notebooks that have extremely tight board space and power constraints. The MAX40000/MAX40001 are available in a 6-bump wafer-level package (WLP) with 1.11mm x 0.76mm footprint and a 6-pin SOT23 package. The MAX40000 has a push-pull output and the MAX40001 has an open-drain output.

The devices offer a supply voltage range from 1.7V to 5.5V and consume only  $0.9\mu\text{A}$  of supply current. They also feature internal filtering to provide high RF immunity, important in many portable applications.

The devices have a high-precision integrated reference that is factory trimmed to an initial accuracy of 1% and better than 2.5% over the entire temperature range. Internal reference voltage options include 1.252V, 1.66V, 1.94V, and 2.22V. See <u>Ordering Information</u> for help with ordering a MAX40000/MAX40001 with a particular voltage reference value and package type. The reference output is stable for capacitive loads up to 100pF.

These devices are fully specified over -40°C to +125°C automotive temperature range.

### **Applications**

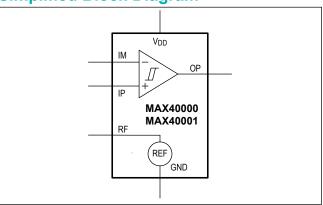
- Cell Phones
- Tablets and Consumer Accessories
- Notebook Computers
- Electronic Toys
- Portable Medical Instruments/Wearables
- Level Detectors

#### **Benefits and Features**

- Micropower Operating Current (0.9μA typ, 1.7μA max)
   Preserves Battery Power
- Tiny 1.11mm x 0.76mm 6-Bump WLP and SOT23 Packages Save Board Space
- Internal Precision Reference Saves Space and Cost of an External Reference
  - < 1% at Room Temperature, < 2.5% Over Temp Reference
  - Multiple Reference Voltages (1.252V, 1.66V, 1.94V, and 2.22V)
- Input Voltage Range = -0.2V to 5.7V
- Supply Voltage Range (1.7V to 5.5V) Allows Operation from 1.8V, 2.5V, 3V, and 5V Supplies
- <10µs Propagation Delay</li>
- Push-Pull (MAX40000) or Open-Drain (MAX40001) Outputs

Ordering Information appears at end of data sheet.

### Simplified Block Diagram





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## MAX40000/MAX40001

# 1.7V, nanoPower Comparators with Built-In Reference

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MAX40000/MAX40001

1.7V, nanoPower Comparators with Built-In

## **Absolute Maximum Ratings**

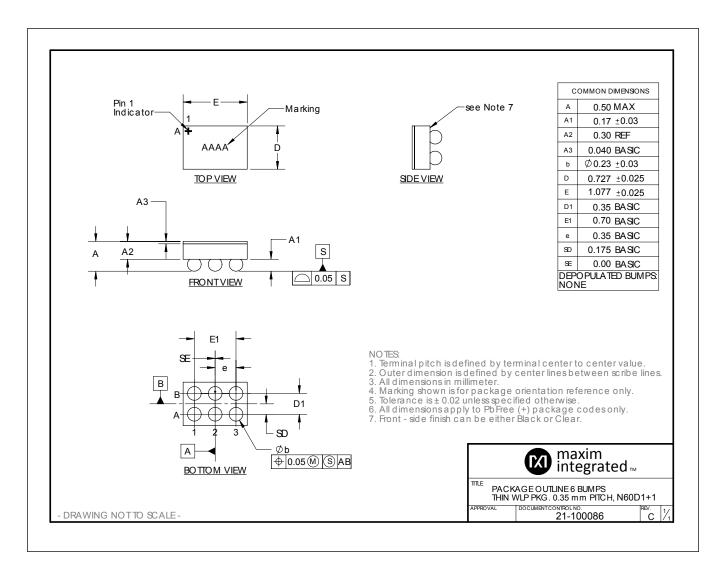
V <sub>DD</sub> to GND	0.3V to +6V	Continuous Power Di
REF to GND	0.3V to +6V	Dissipation (WLP, der
IP, IM to GND	0.3V to +6V	Maximum Power Diss
IP to IM	±6V	+70°C)
OUT (open-drain) to GND	0.3V to +6V	Operating Temperatur
OUT (push-pull) to GND	3V to V <sub>DD</sub> + 0.3V	Junction Temperature
Output Short-Circuit Duration	10s	Storage Temperature
Continuous Current Into Any Input Pin		Lead Temperature (so
Continuous Current Into/Out of Any Output Pin	50mA	Soldering Temperatur

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **Package Information**

#### 6 WLP

Package Code	N60D1+1			
Outline Number	<u>21-100086</u>			
Land Pattern Number	Refer to Application Note 1891			
THERMAL RESISTANCE, FOUR-LAYER BOARD				
Junction to Ambient (θ <sub>JA</sub> )	98°C/W			



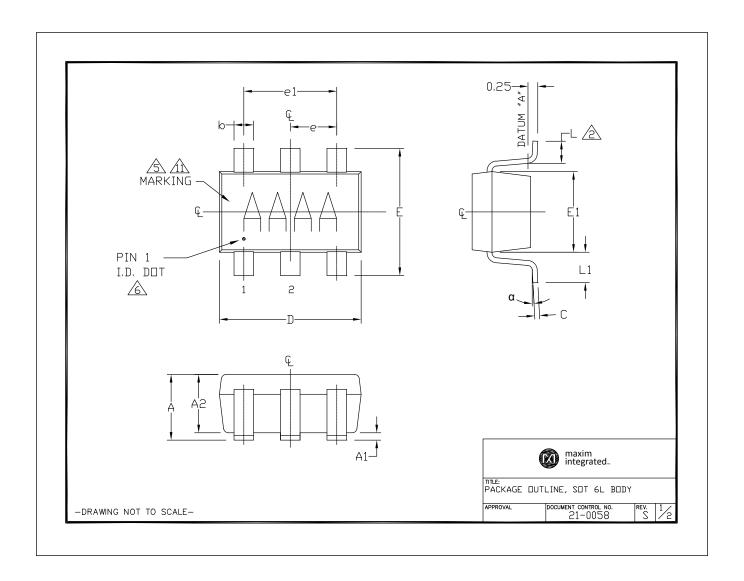
#### **6 SOT23**

Package Code	U6+1				
Outline Number	21-0058				
Land Pattern Number	<u>90-0175</u>				
THERMAL RESISTANCE, FOUR-LAYER BOARD					
Junction to Ambient (θ <sub>JA</sub> )	230°C/W				
Junction to Case (θ <sub>JC</sub> )	76°C/W				

For the latest package outline information and land patterns (footprints), go to <a href="https://www.maximintegrated.com/packages">www.maximintegrated.com/packages</a>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

#### Note:

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to <a href="https://www.maximintegrated.com/thermal-tutorial">www.maximintegrated.com/thermal-tutorial</a>.



#### NOTES:

1. ALL DIMENSIONS ARE IN MILLIMETERS UNLESS OTHER SPECIFIED.

ÉS FOOT LENGTH MEASURED AT INTERCEPT POINT BETWEEN DATUM A & LEAD SURFACE.

- 3. PACKAGE DUTLINE EXCLUSIVE OF MOLD FLASH & METAL BURR, MOLD FLASH, PROTRUSION OR METAL BURR SHOULD NOT EXCEED 0.25mm.
- 4. PACKAGE DUTLINE INCLUSIVE DF SOLDER PLATING.

S PIN 1 IS LOWER LEFT PIN WHEN READING TOP MARK FROM LEFT TO RIGHT.

APIN 1 I.D. DOT IS Ø0.3mm MIN. LOCATED ABOVE PIN 1.

- 7. MEETS JEDEC MO178, VARIATION AB.
- 8. SOLDER THICKNESS MEASURED AT FLAT SECTION OF LEAD BETWEEN 0.08mm AND 0.15mm FROM LEADTIP.
- 9. LEAD TO BE COPLANAR WITHIN 0.1mm.
- 10. NUMBER OF LEADS SHOWN ARE FOR REFERENCE ONLY.

MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.

- 12. MATERIAL MUST BE COMPLIANT WITH MAXIM SPECIFICATION 10-0131 FOR SUBSTANCE CONTENT, MUST BE EU ROHS COMPLIANT WITHOUT EXEMPTION AND PB-FREE.
- 13. ALL DIMENSIONS APPLY TO BOTH LEADED (-) AND PWFREE (+) PKG. CODES

SYMBOL	MIN NOMINAL		MAX		
Α	0.90	1.25	1.45		
A1	0.00	0.05	0.15		
A2	0.90	1.10	1.30		
b	0.35	0.40	0.50		
С	0.08	0.15	0.20		
D	2.80	2.90	3.00		
E	2.60	2.60 2.80			
E1	1.50	1.50 1.625			
L	0.35	0.45	0.60		
L1		0.60 REF.			
e1	1.90 BSC.				
е	0.95 BSC.				
α	0°	2.5*	10°		

PKG CDDES: U6-1, U6-1A, U6-2, U6-4, U6-4A, U6-5, U6-5A, U6CN-2, U6F-6, U6SN-1, U6-8, U6-9, U6-1C

\*\* U6-9 TO BE USED FOR RF50 PARTS ONLY WHICH USES A SI SPACER \*\* U6-5 USES LOW STRESS MOLD COMPOUND



| TITLE: | PACKAGE DUTLINE, SDT 6L BDDY

-DRAWING NOT TO SCALE-

DOCUMENT CONTROL NO. REV. 2

### **Electrical Characteristics**

 $(V_{DD}$  = 3.3V,  $V_{CM}$  = 0V,  $R_{PULLUP}$  = 100k $\Omega$  from OUT to  $V_{PULLUP}$  = 3.3V (for MAX40001 only),  $T_A$  =  $T_{MIN}$  to  $T_{MAX}$ . Typical values are at  $T_A$  = +25°C, unless otherwise noted.) (<u>Note 1</u>)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS	
POWER SUPPLY VOLTAGE							
V <sub>DD</sub> Supply Voltage Range V <sub>DD</sub>	V <sub>REF</sub> < 1.8V, guaranteed by PSRR specification	1.7		5.5	V		
	V <sub>REF</sub> > 1.8V, guaranteed by PSRR specification	V <sub>REF</sub> + 0.1		5.5	V		
V <sub>DD</sub> Supply Current	I <sub>DD</sub>	No output or reference load current, T <sub>A</sub> = -40°C to +125°C		0.9	1.7	μΑ	
Power-Up Time				5		μs	

## **Electrical Characteristics (continued)**

 $(V_{DD}=3.3V,V_{CM}=0V,R_{PULLUP}=100k\Omega \text{ from OUT to }V_{PULLUP}=3.3V \text{ (for MAX40001 only)}, T_{A}=T_{MIN} \text{ to }T_{MAX}. \text{ Typical values are at }T_{A}=+25^{\circ}\text{C}, \text{ unless otherwise noted.)} \text{ ($\underline{Note\ 1}$)}$ 

PARAMETER	SYMBOL	С	ONDITIONS	MIN	TYP	MAX	UNITS
COMPARATOR			1				
Input Common-Mode	V <sub>CM</sub>	T <sub>A</sub> = +25°C		-0.2		V <sub>DD</sub> + 0.2 V	V
Voltage Range	0	$T_A = -40^{\circ}C \text{ to } +$	-125°C	0		$V_{DD}$	
			V <sub>CM</sub> = 0V to V <sub>DD</sub> -1V			8	
Input Offset Voltage	V <sub>OS</sub>	Note 2	$V_{CM} = V_{DD}$ -1V to $V_{DD}$ , $T_A = 0$ °C to +85°C			10	mV
			$V_{CM} = V_{DD} - 1V$ to $V_{DD}$ , $T_A = -40$ °C to $+125$ °C			14	
Input Offset Drift					27		μV/°C
Input Hysteresis	V <sub>HYS</sub>	Note 3			2.5		mV
			V <sub>CM</sub> = -0.2V to V <sub>DD</sub> +0.2V, T <sub>A</sub> = +25°C		1	5	
Input Bias Current	Note 4	$V_{CM} = 0V \text{ to } V_{DD},$ $T_A = -40^{\circ}\text{C to}$ $+85^{\circ}\text{C}$		1	5	nA	
			V <sub>CM</sub> = 0.2V to V <sub>DD</sub> , T <sub>A</sub> = -40°C to +125°C		1	5	
Input Offset Current		Note 4				5	nA
Input Capacitance		Either input, over	er V <sub>CM</sub> range		1.5		pF
Power Supply Rejection Ratio	PSRR	DC, over the er voltage range	DC, over the entire common-mode input voltage range				dB
Common-Mode Rejection Ratio	CMRR	DC, over the er voltage range	ntire common-mode input	46			dB
Output Voltage Swing Low	V <sub>OL</sub>	Sinking 2mA ou V <sub>GND</sub>	utput current, V <sub>OUT</sub> -			0.4	V
Output Voltage Swing High	V <sub>OH</sub>	Sourcing 2mA o	output current,			0.4	V
Output Leakage Current	I <sub>O-LKG</sub>		y (MAX40001), V <sub>DD</sub> = V, T <sub>A</sub> = -40°C to +125°C			100	nA

## **Electrical Characteristics (continued)**

 $(V_{DD}=3.3V,V_{CM}=0V,R_{PULLUP}=100k\Omega \text{ from OUT to }V_{PULLUP}=3.3V \text{ (for MAX40001 only)}, T_{A}=T_{MIN} \text{ to }T_{MAX}. \text{ Typical values are at }T_{A}=+25^{\circ}\text{C}, \text{ unless otherwise noted.)} \text{ ($\underline{Note\ 1}$)}$ 

PARAMETER	SYMBOL	CONI	DITIONS	MIN	TYP	MAX	UNITS		
			Output L->H, MAX40000		9.6				
		100mV overdrive (Note 5)	Output L->H, MAX40001, 100kΩ		14				
			Output H->L, MAX40000		3.2				
Propagation Delay	t <sub>PD</sub>		Output L->H, MAX40000		9.9		μs		
		20mV overdrive (Note 5)	Output L->H, MAX40001, 100kΩ		14.8				
			Output H->L, MAX40000		5.2				
Rise Time	t <sub>R</sub>	Push-pull output sta	age. 25% to 75%		300		ns		
Fall Time	t <sub>F</sub>	25% to 75%			52		ns		
INTERNAL REFERENCE	VOLTAGE								
			MAX40000ANT12+ T	1.2395	1.252	1.2645			
	Forence Voltage	T	MAX4000 _16+		1.66				
		T <sub>A</sub> = +25°C	MAX4000 _19+		1.94				
			MAX4000 _22+		2.22				
		T <sub>A</sub> = -40°C to +85°C			MAX4000 _12+		1.252		
			MAX4000 _16+		1.66		v		
Reference voltage			MAX4000 _19+		1.94				
			MAX4000 _22+		2.22				
			MAX40000ANT12+ T	1.2207	1.252	1.2833			
	$T_{\Lambda} = -40^{\circ} \text{C to}$	T <sub>A</sub> = -40°C to	MAX4000 _16+		1.66				
		+125°C	MAX4000 _19+		1.94				
			MAX4000 _22+		2.22				
Reference Thermal Drift	V <sub>REF-</sub> TEMPCO	Over extended tem T <sub>A</sub> = -40°C to +125	perature range, °C		15		ppm/°C		
Line Regulation						1200	ppm/V		
Load Regulation		I <sub>VREFOUT</sub> = ±100n/	4			0.01	mV/nA		

### **Electrical Characteristics (continued)**

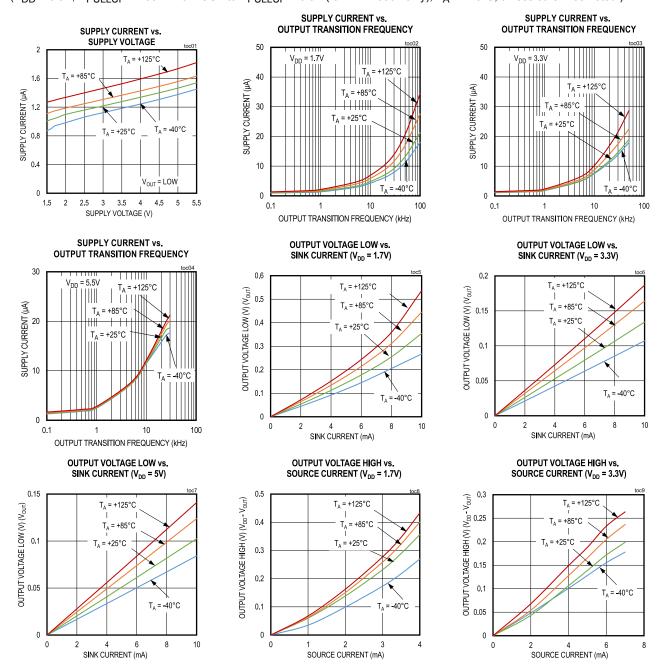
 $(V_{DD}$  = 3.3V,  $V_{CM}$  = 0V,  $R_{PULLUP}$  = 100k $\Omega$  from OUT to  $V_{PULLUP}$  = 3.3V (for MAX40001 only),  $T_A$  =  $T_{MIN}$  to  $T_{MAX}$ . Typical values are at  $T_A$  = +25°C, unless otherwise noted.) (<u>Note 1</u>)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Current			0.1			μA
Voltage Noise		0.1Hz to 10Hz		82		μV <sub>P-P</sub>
Voltage Noise Density		C <sub>REF</sub> = 1nF, 10Hz to 6kHz		2.19		μV/√Hz
Capacitive Load Stability					100	pF

- Note 1: All specifications are 100% production tested at T<sub>A</sub> = +25°C. Specification limits over temperature (T<sub>A</sub> = T<sub>MIN</sub> to T<sub>MAX</sub>) are guaranteed by design, not production tested.
- Note 2: Input offset voltage; VOS is defined as the center of the hysteresis band or average of the threshold trip points.
- Note 3: The hysteresis-related trip points are defined as the edges of the hysteresis band, measured with respect to the center of the band (i.e., Vos) (Figure 1).
- Note 4: Guaranteed by design and characterization.
- Note 5: Specified with an input overdrive (V<sub>OVERDRIVE</sub>) of 100mV and 20mV, and load capacitance of C<sub>L</sub> = 15pF. V<sub>OVERDRIVE</sub> is defined above the offset voltage and hysteresis of the comparator input. For the MAX40000/MAX40001, reference voltage error should also be added.

## **Typical Operating Characteristics**

 $(V_{DD} = 3.3V, R_{PULLUP} = 100 k\Omega \text{ from OUT to } V_{PULLUP} = 3.3V \text{ (for MAX40001 only)}, T_{A} = +25^{\circ}C, \text{ unless otherwise noted.)}$ 

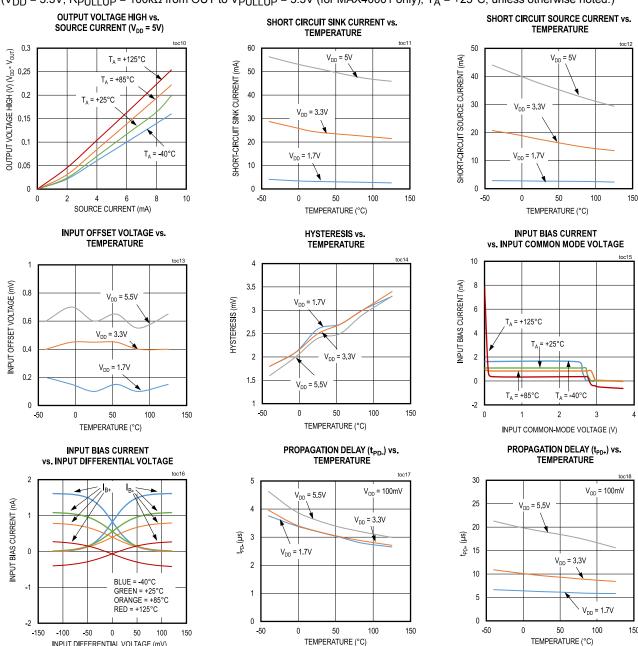


INPUT DIFFERENTIAL VOLTAGE (mV)

## 1.7V, nanoPower Comparators with Built-In Reference

## **Typical Operating Characteristics (continued)**

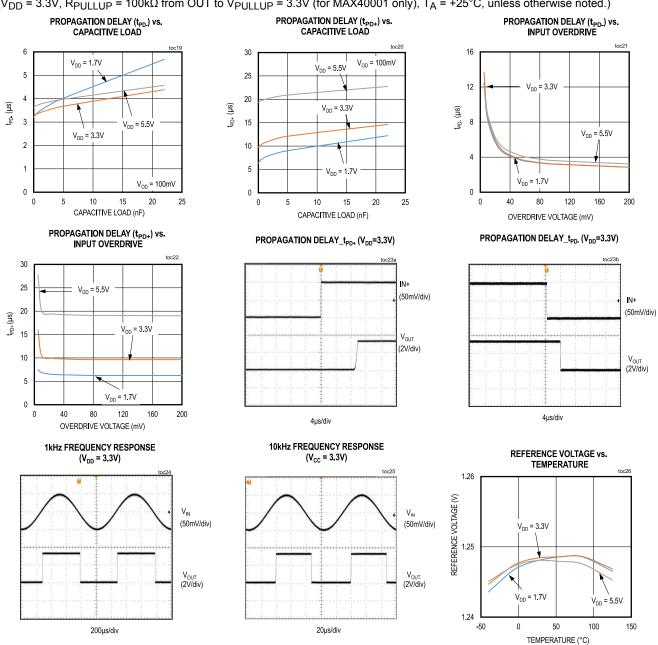
 $(V_{DD} = 3.3V, R_{PULLUP} = 100k\Omega$  from OUT to  $V_{PULLUP} = 3.3V$  (for MAX40001 only),  $T_A = +25^{\circ}C$ , unless otherwise noted.)



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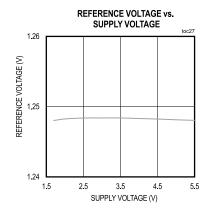
## **Typical Operating Characteristics (continued)**

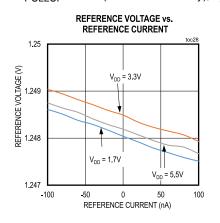
 $(V_{DD} = 3.3V, R_{PULLUP} = 100 k\Omega \text{ from OUT to } V_{PULLUP} = 3.3V \text{ (for MAX40001 only)}, T_{A} = +25 ^{\circ}\text{C, unless otherwise noted.)}$ 

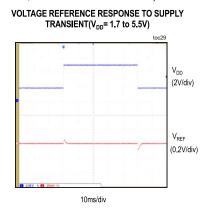


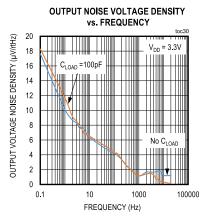
## **Typical Operating Characteristics (continued)**

 $(V_{DD} = 3.3V, R_{PULLUP} = 100kΩ$  from OUT to  $V_{PULLUP} = 3.3V$  (for MAX40001 only),  $T_A = +25$ °C, unless otherwise noted.)



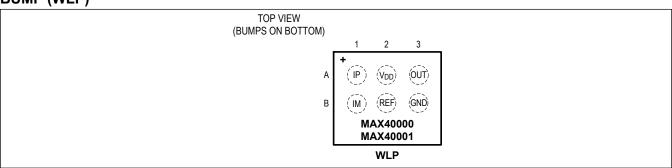




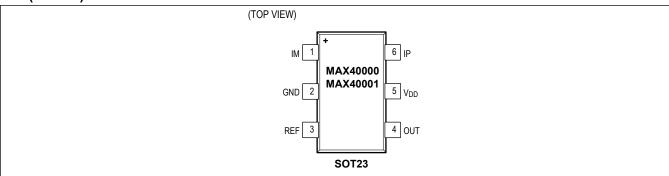


## **Pin Configurations**

### **BUMP (WLP)**



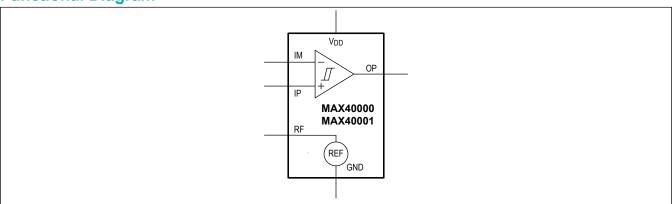
## PIN (SOT23)



## **Pin Description**

P	IN	NAME	FUNCTION
BUMP (WLP)	PIN (SOT23)	NAIVIE	FUNCTION
B1	1	IM	Inverting Input of Comparator
B2	3	REF	Internal Voltage Reference Output. Bypass REF pin with a 0.1µF capacitor to GND as close as possible to the device.
В3	2	GND	Ground
A1	6	IP	Noninverting Input of Comparator
A2	5	V <sub>DD</sub>	$V_{DD}$ Supply Voltage. Bypass $V_{DD}$ with a 0.1 $\mu$ F capacitor to GND as close as possible to the device pin.
A3	4	OUT	Open-Drain Output (MAX40001)/Push-Pull Output (MAX40000). For the open-drain version, connect a $100k\Omega$ pullup resistor from OUT to any pullup voltage up to $5.5V$ .

## **Functional Diagram**



### **Detailed Description**

The MAX40000/MAX40001 feature an on-board voltage reference with ±1% initial accuracy. This family of comparators with internal references are available in multiple voltage reference options. The <u>Ordering Information</u> table provides exact part numbers associated with a particular voltage reference option. The common-mode voltage range of this family extends 200mV beyond the rails, allowing signals slightly beyond the rails to trigger the comparator. The 2.5mV internal hysteresis ensures clean output switching even with slow moving input signals. Large internal output drivers allow rail-to-rail output swing with up to ±2mA loads.

The output stage employs a unique design that minimizes supply current surges while switching, virtually eliminating supply glitches typical of many other comparators. The MAX40000 has a push-pull output stage that sinks as well as sources current. The MAX40001 has an open-drain output stage that can be pulled beyond  $V_{DD}$  to a maximum of 5.5V above GND. Multiple comparators with open-drain outputs (OUT) can be connected together in parallel and share a single pullup resistor. This enables user to detect if there is any fault if at least one comparator trips different to other comparators.

#### **Input Stage Circuitry**

The input common-mode voltage range extends from - 0.2V to  $V_{DD}$  + 0.2V. These comparators operate at any differential input voltage within these limits. Input bias current is typically  $\pm 1nA$  if the input voltage is between the supply rails.

#### **Output Stage Structure**

The devices contain a unique break-before-make output stage capable of rail-to-rail operation with up to ±2mA loads. Many comparators consume orders of magnitude more current during switching than during steady-state operation. In the <u>Typical Operating Characteristics</u>, the Supply Current graphs show the minimal supply-current increase as the output switching frequency approaches 1kHz. This characteristic reduces the need for power-supply filter capacitors to reduce glitches created by comparator switching currents. In battery-powered applications, this characteristic results in a substantial increase in battery life.

#### **Voltage Reference**

The MAX40000/MAX40001 come with different internal voltage reference options that has initial accuracy of  $\pm 1\%$ . 1.252V, 1.6V, 1.9V, and 2.2V options of internal voltage references are available. The devices' internal reference has a typical temperature coefficient of 15ppm/°C over the full -40°C to +125°C temperature range. The reference is a very-low-power bandgap cell, with a maximum 10k $\Omega$  output impedance. REF pin can source and sink up to 100nA to external circuitry. For applications that need increased drive, buffer REF with a low input-bias current op amp such as the MAX44265. Most applications require no bypass capacitor on REF pin.

### **Applications Information**

### **Battery-Powered Operation**

The MAX40000 and MAX40001 are ideally suited for use with most battery-powered systems. <u>Table 1</u> lists Alkaline and Lithium-Ion batteries with capacities and approximate operating times for MAX40000 and MAX40001, assuming nominal conditions.

#### **Internal Hysteresis**

Many comparators oscillate in the linear region of operation because of noise or undesired parasitic feedback. This tends to occur when the voltage on one input is equal or very close to the voltage on the other input. The MAX40000/MAX40001 have internal 2.5mV hysteresis to counter parasitic effects and noise.

The hysteresis in a comparator creates two trip points: one for upper threshold ( $V_{TRIP+}$ ) and one for lower threshold ( $V_{TRIP-}$ ) for voltage transitions on the input signal (<u>Figure 1</u>). The difference between the trip points is the hysteresis band ( $V_{HYS}$ ). When the comparator's input voltages are equal, the hysteresis effectively causes one comparator input to move quickly past the other, thus taking the input out of the region where oscillation occurs. <u>Figure 1</u> illustrates the case in which IM has a fixed voltage applied, and IP is varied. If the inputs were reversed, the figure would be the same, except with an inverted output.

#### **Adding External Hysteresis**

In applications requiring more than the internal 2.5mV hysteresis of the devices, additional hysteresis can be added with two external resistors. Since these comparators are intended to use in very low-power systems, care must be taken to minimize power dissipation in the additional circuitry.

Regardless of which approach is employed to add external hysteresis, the external hysteresis will be  $V_{DD}$  dependent. Over the full discharge range of battery-powered systems, the hysteresis can change as much as 40%. <u>Figure 2</u> shown below is simplest circuit for adding external hysteresis. In this example, the hysteresis is defined by:

Hysteresis = 
$$\frac{R_G}{R_F} \times V_{DD}$$

Where  $R_G$  is the source resistance and  $R_F$  is the feedback resistance. Because the comparison threshold is 1/2  $V_{DD}$ , the MAX40000 was chosen for its push-pull output and lack of reference. This provides symmetrical hysteresis around the threshold.

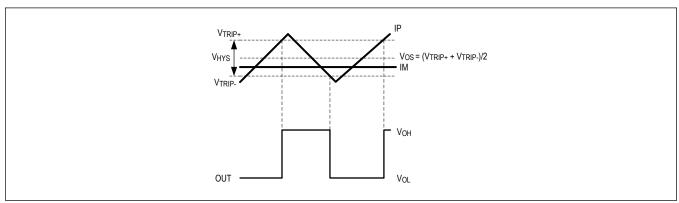


Figure 1. Hysteresis Band

<b>Table 1. Battery</b>	<b>Applications</b>	Using MAX40000	and MAX40001

BATTERY TYPE	RECHARGEABLE	V <sub>INITIAL</sub> (V)	V <sub>END-OF-</sub> LIFE (V)	CAPACITY, AA SIZE (mAh)	MAX40000/MAX40001 OPERATING TIME (hr)
Alkaline (2 Cells)	No	3.0	1.8	2000	1.8 x 10 <sup>6</sup>
Lithium-Ion (1 Cell)	Yes	3.5	2.7	1000	0.9 x 10 <sup>6</sup>

### **Output Considerations**

In most cases, the push-pull output of MAX40000 is best for external hysteresis. The open-drain output of the MAX40001 can be used, but the effect of the feedback network and pullup resistor on the actual output high voltage must be considered.

#### **Component Selection**

Because the MAX40000/MAX40001 are intended for very low power-supply systems, the highest impedance circuits should be used wherever possible. The offset error due to input-bias current is proportional to the total impedance seen at the input. For example, selecting components for Figure 2, with a target of 50mV hysteresis, a 5V supply, and choosing an  $R_F$  of  $10M\Omega$  gives  $R_G$  as  $100k\Omega$ . The total impedance seen at IN+ is therefore  $10M\Omega \parallel 100k\Omega$ , or  $99k\Omega$ . The maximum input bias current of MAX40000/MAX40001 is 1nA; therefore, the error due to source impedance is less than  $100\mu V$ .

#### **Board Layout and Bypassing**

Power-supply bypass capacitors are not typically needed, but use 100nF bypass capacitors close to the device's supply pins when supply impedance is high, supply leads are long, or excessive noise is expected on the supply lines. Minimize signal trace lengths to reduce stray capacitance. A ground plane and surface-mount components are recommended. If the REF pin is decoupled, use a new low-leakage capacitor.

#### **Logic-Level Translator**

The <u>Typical Application Circuit</u> shows an application that converts 5V logic to 3V logic levels. The MAX40001 is powered by the +5V supply voltage to  $V_{DD}$ , and the pullup resistor for the MAX40001's open-drain output is connected to the +3V supply voltage. This configuration allows the full 5V logic swing without creating overvoltage on the 3V logic inputs. For 3V to 5V logic-level translations, simply connect the +3V supply voltage to  $V_{DD}$  and the +5V supply voltage to the pullup resistor.

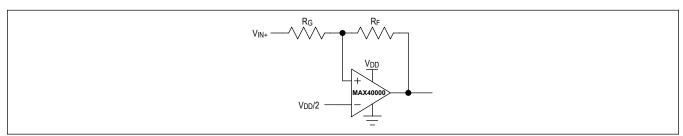
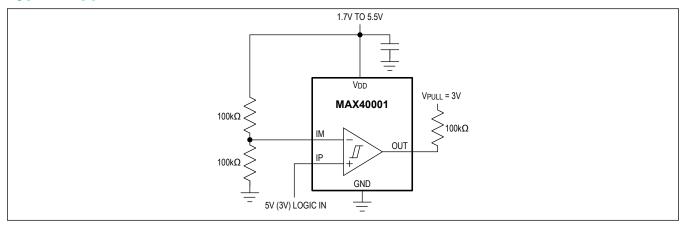


Figure 2. External Hysteresis on MAX40000

## **Typical Application Circuit**



## **Ordering Information**

PART	TEMP RANGE	PIN- PACKAGE	TOP MARK
MAX40000ANT12+T	-40°C to +125°C	6 WLP	+N
MAX40000ANT16+T*	-40°C to +125°C	6 WLP	+0
MAX40000ANT19+T*	-40°C to +125°C	6 WLP	+P
MAX40000ANT22+T*	-40°C to +125°C	6 WLP	+Q
MAX40000AUT12+T	-40°C to +125°C	6 SOT23	_
MAX40000AUT16+T*	-40°C to +125°C	6 SOT23	_
MAX40000AUT19+T*	-40°C to +125°C	6 SOT23	_
MAX40000AUT22+T*	-40°C to +125°C	6 SOT23	-
MAX40001ANT12+T	-40°C to +125°C	6 WLP	+R
MAX40001ANT16+T*	-40°C to +125°C	6 WLP	+S
MAX40001ANT19+T*	-40°C to +125°C	6 WLP	+T
MAX40001ANT22+T	-40°C to +125°C	6 WLP	+U
MAX40001AUT12+T	-40°C to +125°C	6 SOT23	-
MAX40001AUT16+T*	-40°C to +125°C	6 SOT23	_
MAX40001AUT19+T*	-40°C to +125°C	6 SOT23	
MAX40001AUT22+T	-40°C to +125°C	6 SOT23	_

<sup>+</sup>Denotes a lead (Pb)-free/RoHS-compliant package.

For example, the MAX40000ANT12+T has an onboard 1.2V reference voltage.

Devices without "\_\_" use external reference voltage as supply voltage.

T = Tape and reel.

<sup>\*</sup>Future product—contact factory for availability.

## MAX40000/MAX40001

# 1.7V, nanoPower Comparators with Built-In Reference

## **Revision History**

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	1/17	Initial release	_
1	3/17	Updated title to include "nanoPower"	1–14
2	4/17	Updated Ordering Information table	13
3	5/17	Updated Ordering Information table	13
4	8/17	Updated Functional Diagram and Ordering Information table	10, 13
5	1/21	Added package outline drawings	6–8
6	5/21	Updated Electrical Characteristics table	9–10

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